

SLUS763C-JULY 2007-REVISED OCTOBER 2007

# OVERVOLTAGE AND OVERCURRENT PROTECTION IC AND Li+ CHARGER FRONT-END PROTECTION IC

# FEATURES

- Provides Protection for Three Variables:
  - Input Overvoltage, with Rapid Response in < 1 μs</li>
  - User-Programmable Overcurrent with Current Limiting
  - Battery Overvoltage
- 30V Maximum Input Voltage
- Supports up to 1.5A Input Current
- Robust Against False Triggering Due to Current Transients
- Thermal Shutdown
- Enable Input
- Status Indication Fault Condition

# DESCRIPTION

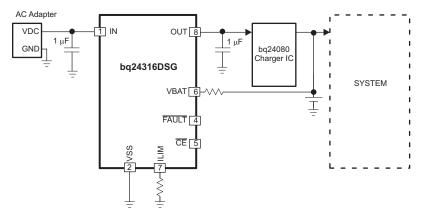
#### • Available in Space-Saving Small 8 Lead 2×2 SON and 12 Lead 4x3 SON Packages

## **APPLICATIONS**

- Mobile Phones and Smart Phones
- PDAs
- MP3 Players
- Low-Power Handheld Devices
- Bluetooth Headsets

The bq24314 and bq24316 are highly integrated circuits designed to provide protection to Li-ion batteries from failures of the charging circuit. The IC continuously monitors the input voltage, the input current, and the battery voltage. In case of an input overvoltage condition, the IC immediately removes power from the charging circuit by turning off an internal switch. In the case of an overcurrent condition, it limits the system current at the threshold value, and if the overcurrent persists, switches the pass element OFF after a blanking period. Additionally, the IC also monitors its own die temperature and switches off if it becomes too hot. The input overcurrent threshold is user-programmable.

The IC can be controlled by a processor and also provides status information about fault conditions to the host.



# **APPLICATION SCHEMATIC**

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet. PowerPAD is a trademark of Texas Instruments.







These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

#### ORDERING INFORMATION<sup>(1)</sup>

DEVICE <sup>(2)</sup>	OVP THRESHOLD	PACKAGE	MARKING
bq24314DSG	5.85 V	2mm x 2mm SON	CBV
bq24314DSJ	5.85 V	4mm x 3mm SON	CBX
bq24316DSG	6.80 V	2mm x 2mm SON	CBW
bq24316DSJ	6.80 V	4mm x 3mm SON	BZC

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

(2) To order a 3000 pcs reel add R to the part number, or to order a 250 pcs reel add T to the part number.

### PACKAGE DISSIPATION RATINGS

PART NO.	PACKAGE	R <sub>θJC</sub>	R <sub>0JA</sub>
BQ24314DSG BQ24316DSG	2×2 SON	5°C/W	75°C/W
BQ24314DSJ BQ24316DSJ	4×3 SON	5°C/W	40°C/W

### **ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>**

over operating free-air temperature range (unless otherwise noted)

PARAMETER	PIN	VALUE	UNIT
	IN (with respect to VSS)	-0.3 to 30	
Input voltage	OUT (with respect to VSS)	-0.3 to 12	V
	ILIM, FAULT, CE, VBAT (with respect to VSS)	-0.3 to 7	-
Input current	IN	2.0	А
Output current	OUT	2.0	А
Output sink current	FAULT	15	mA
Junction temperature, T <sub>J</sub>		-40 to 150	°C
Storage temperature, T <sub>STG</sub>		-65 to 150	°C
Lead temperature (soldering, 10 seconds)		300	°C

(1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. All voltage values are with respect to the network ground terminal unless otherwise noted.

# **RECOMMENDED OPERATING CONDITIONS**

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V <sub>IN</sub>	Input voltage range	3.3	26	V
I <sub>IN</sub>	Input current, IN pin		1.5	А
I <sub>OUT</sub>	Output current, OUT pin		1.5	А
R <sub>ILIM</sub>	OCP Programming resistor	15.0	90.0	kΩ
TJ	Junction temperature	0	125	°C



# **ELECTRICAL CHARACTERISTICS**

over junction temperature range  $0^{\circ}C \le T_{J} \le 125^{\circ}C$  and recommended supply voltage (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
IN			1				
V <sub>UVLO</sub>	Under-voltage lock- power detected thre		$\overline{CE}$ = Low, V <sub>IN</sub> increasing from 0V to 3V	2.6	2.7	2.8	V
V <sub>HYS-UVLO</sub>	Hysteresis on UVLC	)	$\overline{CE}$ = Low, V <sub>IN</sub> decreasing from 3V to 0V	200	260	300	mV
T <sub>DGL(PGOOD)</sub>	Deglitch time, input detected status	power	$\overline{CE}$ = Low. Time measured from V_IN 0V $\rightarrow$ 5V 1µs rise-time, to output turning ON		8		ms
I <sub>DD</sub>	Operating current		$\overline{\text{CE}}$ = Low, No load on OUT pin, V <sub>IN</sub> = 5V, R <sub>ILIM</sub> = 25k $\Omega$		400	600	μA
ISTDBY	Standby current		$\overline{CE}$ = High, V <sub>IN</sub> = 5.0V		65	95	μA
INPUT TO O	UTPUT CHARACTE	RISTICS					
VDO	Drop-out voltage IN	to OUT	$\overline{CE}$ = Low, V <sub>IN</sub> = 5V, I <sub>OUT</sub> = 1A		170	280	mV
INPUT OVER	RVOLTAGE PROTEC	CTION					
	Input overvoltage	bq24314		5.71	5.85	6.00	V
V <sub>OVP</sub>	protection threshold	bq24316	$\overline{CE}$ = Low, V <sub>IN</sub> increasing from 5V to 7.5V	6.60	6.80	7.00	V
t <sub>PD(OVP)</sub>	Input OV propagation	on delay <sup>(1)</sup>	CE = Low			1	μs
V <sub>HYS-OVP</sub>	Hysteresis on OVP		$\overline{CE}$ = Low, V <sub>IN</sub> decreasing from 7.5V to 5V	25	60	110	mV
t <sub>ON(OVP)</sub>	Recovery time from overvoltage condition		$\overline{\text{CE}}$ = Low, Time measured from $V_{\text{IN}}$ 7.5V $\rightarrow$ 5V, 1µs fall-time		8		ms
INPUT OVER	RCURRENT PROTEC	CTION					
I <sub>OCP</sub>	Input overcurrent pr threshold range	otection		300		1500	mA
I <sub>OCP</sub>	Input overcurrent pr threshold	otection	$\overline{\text{CE}}$ = Low, R <sub>ILIM</sub> = 25k $\Omega$	930	1000	1070	mA
t <sub>BLANK(OCP)</sub>	Blanking time, input detected	overcurrent			176		μs
t <sub>REC(OCP)</sub>	Recovery time from overcurrent conditio	•			64		ms
BATTERY O	VERVOLTAGE PRO	TECTION					
BV <sub>OVP</sub>	Battery overvoltage threshold	protection	$\overline{CE} = Low, V_{IN} > 4.4V$	4.30	4.35	4.4	V
V <sub>HYS-BOVP</sub>	Hysteresis on BV <sub>OV</sub>	P	$\overline{CE} = Low, V_{IN} > 4.4V$	200	275	320	mV
	Input bias current	DSG Package	$V_{BAT} = 4.4V, T_J = 25^{\circ}C$			10	
I <sub>VBAT</sub>	on VBAT pin	DSJ Package	$V_{BAT} = 4.4V, T_{J} = 85^{\circ}C$			10	nA
T <sub>DGL(BOVP)</sub>	Deglitch time, batter detected	y overvoltage	$\overline{CE}$ = Low, V <sub>IN</sub> > 4.4V. Time measured from V <sub>VBAT</sub> rising from 4.1V to 4.4V to FAULT going low.		176		μs
THERMAL P	ROTECTION						
T <sub>J(OFF)</sub>	Thermal shutdown t	emperature			140	150	°C
T <sub>J(OFF-HYS)</sub>	Thermal shutdown I	•			20		°C
LOGIC LEVE				L			L
V <sub>IL</sub>	Low-level input volta	age		0		0.4	V
V <sub>IH</sub>	High-level input volt	•		1.4			V
 I <sub>IL</sub>	Low-level input curr	-	V <sub>CE</sub> = 0V			1	μA
I <sub>IH</sub>	High-level input curi		$V_{CE} = 1.8V$			15	μA
	ELS ON FAULT					-	
LOGIC LEVE	-						
V <sub>OL</sub>	Output low voltage		I <sub>SINK</sub> = 5mA			0.2	V

(1) Not tested in production. Specified by design.

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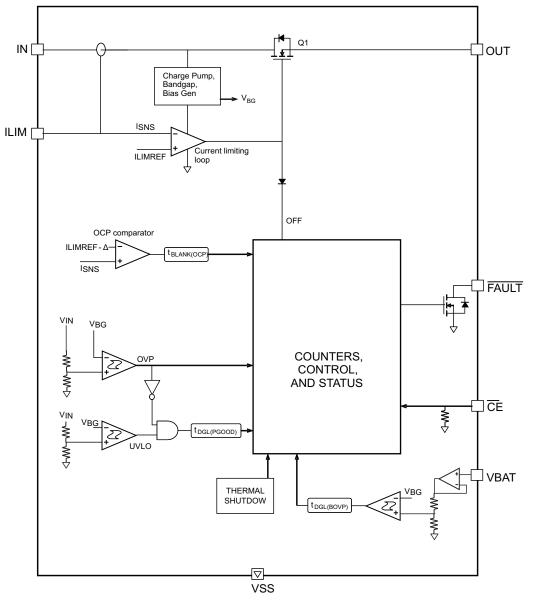
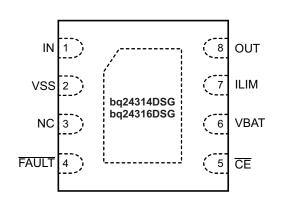
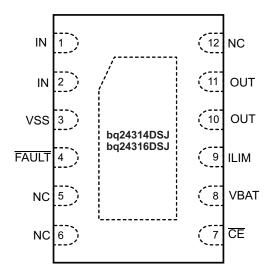


Figure 1. Simplified Block Diagram

## **TERMINAL FUNCTIONS**

TERMINAL		TERMINAL		TERMINAL		TERMINAL		TERMINAL		TERMINAL		TERMINAL		TERMINAL		TERMINAL		ERMINAL		ERMINAL		TERMINAL		TERMINAL						
NAME	DSJ	DSG	I/O	DESCRIPTION																										
IN	1, 2	1	I	Input power, connect to external DC supply. Connect external $1\mu$ F ceramic capacitor (minimum) to VSS. For the 12 pin (DSJ-suffix) device, ensure that pins 1 and 2 are connected together on the PCB at the device.																										
OUT	10, 11	8	0	Output terminal to the charging system. Connect external 1µF ceramic capacitor (minimum) to VSS.																										
VBAT	8	6	Ι	Battery voltage sense input. Connect to pack positive terminal through a resistor.																										
ILIM	9	7	I/O	Input overcurrent threshold programming. Connect a resistor to VSS to set the overcurrent threshold.																										
CE	7	5	I	Chip enable input. Active low. When $\overline{CE}$ = High, the input FET is off. Internally pulled down.																										
FAULT	4	4	0	Open-drain output, device status. FAULT = Low indicates that the input FET Q1 has been turned off due to input overvoltage, input overcurrent, battery overvoltage, or thermal shutdown.																										
VSS	3	2	-	Ground terminal																										
NC	5, 6, 12	3		These pins may have internal circuits used for test purposes. Do not make any external connections at these pins for normal operation.																										
Thermal PAD			_	There is an internal electrical connection between the exposed thermal pad and the VSS pin of the device. The thermal pad must be connected to the same potential as the VSS pin on the printed circuit board. Do not use the thermal pad as the primary ground input for the device. The VSS pin must be connected to ground at all times.																										





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# **TYPICAL OPERATING PERFORMANCE**

Test conditions (unless otherwise noted) for typical operating performance:  $V_{IN} = 5 V$ ,  $C_{IN} = 1 \mu F$ ,  $C_{OUT} = 1 \mu F$ ,  $R_{ILIM} = 25 k\Omega$ ,  $R_{BAT} = 100 k\Omega$ ,  $T_A = 25^{\circ}C$ ,  $V_{PU} = 3.3V$  (see Figure 23 for the Typical Application Circuit)  $v_{\rm IN}$ VIN 1 V<sub>OUT</sub> VOUT FAULT IOUT 3→ 1.68 V 2.00 V M2.00ms Ch1 J Ch1 2.00 V Ch2 5.00 V 5.00 V Ch1 Ch2 2.00 V M2.00ms Ch1 J 2.9 V 500m/ Ch3 Figure 2. Normal Power-On Showing Soft-Start,  $R_{OUT}$  = 6.6 $\Omega$ Figure 3. OVP at Power-On,  $V_{IN} = 0V$  to 9V,  $t_r = 50\mu s$ ∆: 3.48 V @: 6.76 V ∆: 3.56 V @: 6.84 V VIN  $v_{\rm IN}$ Max V<sub>OUT</sub> = 6.84 V Max V<sub>OUT</sub> = 6.76 V V<sub>OUT</sub> VOUT FAULT FAULT 4 2.00 V 5.00 V Ch1 2.00 V M 5.00µs Ch1 J Ch1 2.00 V M 20.0µs Ch1 J 6.60 V 2.00 V 6.60 V Ch2 Ch2 1↓ 1↓ Figure 4. bq24316 OVP Response for Input Step,  $V_{IN} = 5V$  to 12V,  $t_r = 1\mu s$ Figure 5. bq24316 OVP Response for Input Step, V<sub>IN</sub> = 5V to 12V,  $t_r = 20 \mu s$ ∆: 5.08 V @: 5.92 V ∆: 5.00 V @: 5.84 V  $\mathbf{v}_{\mathrm{IN}}$  $v_{\rm IN}$ Max V<sub>OUT</sub> = 6.84 V Max V<sub>OUT</sub> = 6.76 V VOUT V<sub>OUT</sub> FAULT FAULT Ch1 2.00 V 2.00 V M 5.00µs Ch1 J 6.64 V Ch1 2.00 V Ċh2 2.00 V M 20.0µs Ch1 J 6.64 V 1↓ 2↓ Figure 7. bq24314 OVP Response for Input Step, V<sub>IN</sub> = 5V Figure 6. bq24314 OVP Response for Input Step, VIN = 5V to 12V,  $t_r = 1 \mu s$ to 12V,  $t_r = 20\mu s$ 

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# **TYPICAL OPERATING PERFORMANCE (continued)**

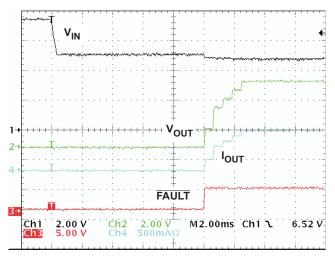
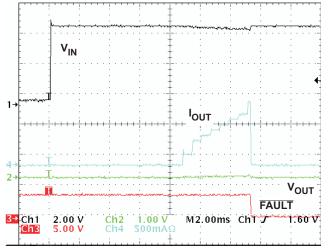
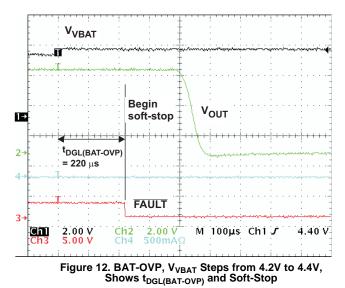


Figure 8. Recovery from OVP,  $V_{IN} = 7.5V$  to 5V,  $t_f = 400 \mu s$ 







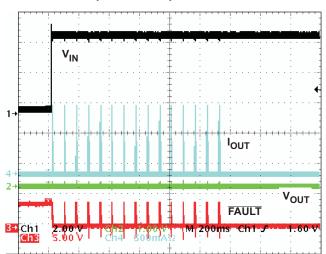


Figure 9. OCP, Powering Up into a Short Circuit on OUT Pin, OCP Counter Counts to 15 Before Switching OFF the Device

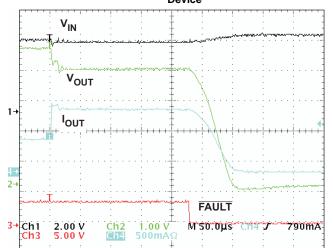


Figure 11. OCP,  $R_{OUT}$  Switches from 6.6 $\Omega$  to 3.3 $\Omega$ , Shows Current Limiting and Soft-Stop

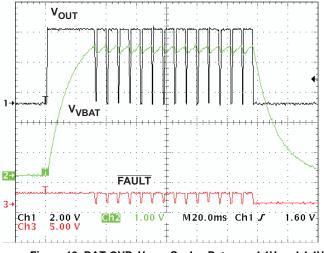
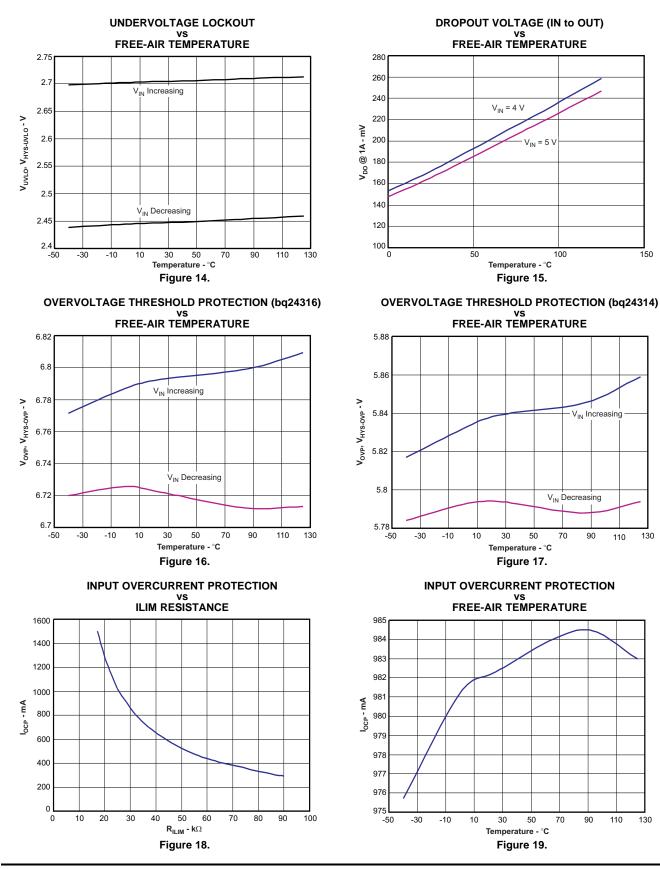


Figure 13. BAT-OVP, V<sub>VBAT</sub> Cycles Between 4.1V and 4.4V, Shows BAT-OVP Counter



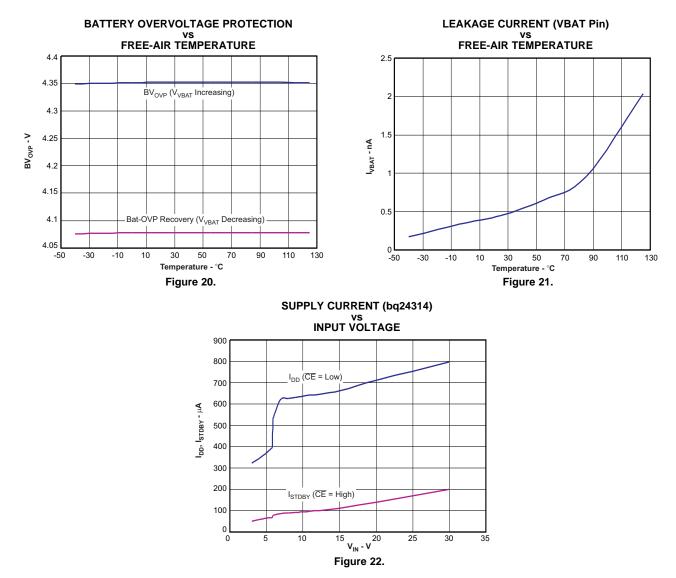
# **TYPICAL OPERATING PERFORMANCE (continued)**



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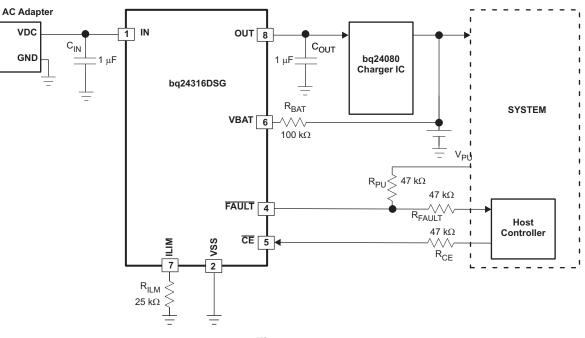


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## TYPICAL APPLICATION CIRCUIT

 $V_{OVP}$  = 6.8V,  $I_{OCP}$  = 1000mA,  $BV_{OVP}$  = 4.35V (Terminal numbers shown are for the 2×2 DSG package)



#### Figure 23.

#### **DETAILED FUNCTIONAL DESCRIPTION**

The bq24314 and bq24316 are highly integrated circuits designed to provide protection to Li-ion batteries from failures of the charging circuit. The IC continuously monitors the input voltage, the input current and the battery voltage. In case of an input overvoltage condition, the IC immediately removes power from the charging circuit by turning off an internal switch. In the case of an overcurrent condition, it limits the system current at the threshold value, and if the overcurrent persists, switches the pass element OFF after a blanking period. If the battery voltage rises to an unsafe level, the IC disconnects power from the charging circuit until the battery voltage returns to an acceptable value. Additionally, the IC also monitors its own die temperature and switches off if it becomes too hot. The input overcurrent threshold is user-programmable. The IC can be controlled by a processor, and also provides status information about fault conditions to the host.

#### POWER DOWN

The device remains in power down mode when the input voltage at the IN pin is below the undervoltage threshold  $V_{UVLO}$ . The FET Q1 connected between IN and OUT pins is off, and the status output, FAULT, is set to Hi-Z.

### **POWER-ON RESET**

The device resets when the input voltage at the IN pin exceeds the UVLO threshold. All internal counters and other circuit blocks are reset. The IC then waits for duration  $t_{DGL(PGOOD)}$  for the input voltage to stabilize. If, after  $t_{DGL(PGOOD)}$ , the input voltage and battery voltage are safe, FET Q1 is turned ON. The IC has a soft-start feature to control the inrush current. The soft-start minimizes the ringing at the input (the ringing occurs because the parasitic inductance of the adapter cable and the input bypass capacitor form a resonant circuit). Figure 2 shows the power-up behavior of the device. Because of the deglitch time at power-on, if the input voltage rises rapidly to beyond the OVP threshold, the device will not switch on at all, instead it will go into protection mode and indicate a fault on the FAULT pin, as shown in Figure 3.



# OPERATION

The device continuously monitors the input voltage, the input current, and the battery voltage as described in detail in the following sections.

#### Input Overvoltage Protection

If the input voltage rises above  $V_{OVP}$ , the internal FET Q1 is turned off, removing power from the circuit. As shown in Figure 4 to Figure 7, the response is very rapid, with the FET turning off in less than a microsecond. The FAULT pin is driven low. When the input voltage returns below  $V_{OVP} - V_{HYS-OVP}$  (but is still above  $V_{UVLO}$ ), the FET Q1 is turned on again after a deglitch time of  $t_{ON(OVP)}$  to ensure that the input supply has stabilized. Figure 8 shows the recovery from input OVP.

#### Input Overcurrent Protection

The overcurrent threshold is programmed by a resistor  $R_{ILIM}$  connected from the ILIM pin to VSS. Figure 18 shows the OCP threshold as a function of  $R_{ILIM}$ , and may be approximated by the following equation:  $I_{OCP} = 25 \div R_{ILIM}$  (current in A, resistance in  $k\Omega$ )

If the load current tries to exceed the  $I_{OCP}$  threshold, the device limits the current for a blanking duration of  $t_{BLANK(OCP)}$ . If the load current returns to less than  $I_{OCP}$  before  $t_{BLANK(OCP)}$  times out, the device continues to operate. However, if the overcurrent situation persists for  $t_{BLANK(OCP)}$ , the FET Q1 is turned off for a duration of  $t_{REC(OCP)}$ , and the FAULT pin is driven low. The FET is then turned on again after  $t_{REC(OCP)}$  and the current is monitored all over again. Each time an OCP fault occurs, an internal counter is incremented. If 15 OCP faults occur in one charge cycle, the FET is turned off permanently. The counter is cleared either by removing and re-applying input power, or by disabling and re-enabling the device with the CE pin. Figure 9 to Figure 11 show what happens in an overcurrent fault.

To prevent the input voltage from spiking up due to the inductance of the input cable, Q1 is turned off slowly, resulting in a "soft-stop", as shown in Figure 11.

#### Battery Overvoltage Protection

The battery overvoltage threshold  $BV_{OVP}$  is internally set to 4.35V. If the battery voltage exceeds the  $BV_{OVP}$  threshold, the FET Q1 is turned off, and the FAULT pin is driven low. The FET is turned back on once the battery voltage drops to  $BV_{OVP} - V_{HYS-BOVP}$  (see Figure 12 and Figure 13). Each time a battery overvoltage fault occurs, an internal counter is incremented. If 15 such faults occur in one charge cycle, the FET is turned off permanently. The counter is cleared either by removing and re-applying input power, or by disabling and re-enabling the device with the  $\overline{CE}$  pin. In the case of a battery overvoltage fault, Q1 is switched OFF gradually (see Figure 12).

#### **Thermal Protection**

If the junction temperature of the device exceeds  $T_{J(OFF)}$ , the FET Q1 is turned off, and the FAULT pin is driven low. The FET is turned back on when the junction temperature falls below  $T_{J(OFF)} - T_{J(OFF-HYS)}$ .

#### Enable Function

The IC has an enable pin which can be used to enable or disable the device. When the  $\overline{CE}$  pin is driven high, the internal FET is turned off. When the  $\overline{CE}$  pin is low, the FET is turned on if other conditions are safe. The OCP counter and the Bat-OVP counter are both reset when the device is disabled and re-enabled. The  $\overline{CE}$  pin has an internal pulldown resistor and can be left floating. Note that the FAULT pin functionality is also disabled when the  $\overline{CE}$  pin is high.

#### Fault Indication

The FAULT pin is an active-low open-drain output. It is in a high-impedance state when operating conditions are safe, or when the device is disabled by setting CE high. With CE low, the FAULT pin goes low whenever any of these events occurs:

- Input overvoltage
- Input overcurrent
- Battery overvoltage
- IC Overtemperature





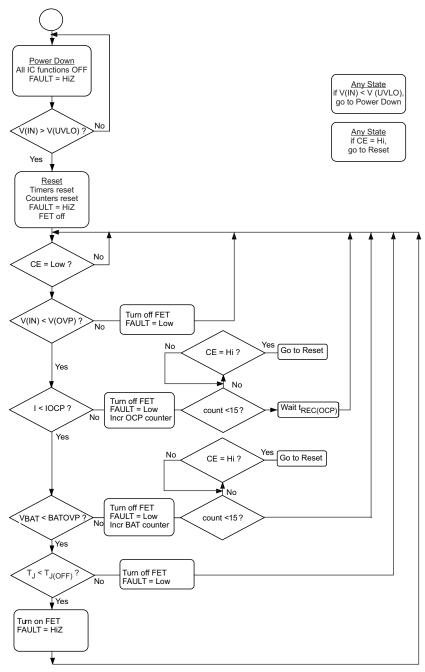


Figure 24. Flow Diagram



# APPLICATION INFORMATION (WITH REFERENCE TO FIGURE 23)

### Selection of R<sub>BAT</sub>

It is strongly recommended that the battery not be tied directly to the VBAT pin of the device, as under some failure modes of the IC, the voltage at the IN pin may appear on the VBAT pin. This voltage can be as high as 30V, and applying 30V to the battery in case of the failure of the bq2431x can be hazardous. Connecting the VBAT pin through  $R_{BAT}$  prevents a large current from flowing into the battery in case of a failure of the IC. In the interests of safety,  $R_{BAT}$  should have a very high value. The problem with a large  $R_{BAT}$  is that the voltage drop across this resistor because of the VBAT bias current  $I_{VBAT}$  causes an error in the BV<sub>OVP</sub> threshold. This error is over and above the tolerance on the nominal 4.35V BV<sub>OVP</sub> threshold.

Choosing  $R_{BAT}$  in the range  $100k\Omega$  to  $470k\Omega$  is a good compromise. In the case of an IC failure, with  $R_{BAT}$  equal to  $100k\Omega$ , the maximum current flowing into the battery would be  $(30V - 3V) \div 100k\Omega = 246\mu$ A, which is low enough to be absorbed by the bias currents of the system components.  $R_{BAT}$  equal to  $100k\Omega$  would result in a worst-case voltage drop of  $R_{BAT} \times I_{VBAT} = 1$ mV. This is negligible to compared to the internal tolerance of 50mV on  $BV_{OVP}$  threshold.

If the Bat-OVP function is not required, the VBAT pin should be connected to VSS.

### Selection of R<sub>CE</sub>, R<sub>FAULT</sub>, and R<sub>PU</sub>

The  $\overline{CE}$  pin can be used to enable and disable the IC. If host control is not required, the  $\overline{CE}$  pin can be tied to ground or left un-connected, permanently enabling the device.

In applications where external control is required, the  $\overline{CE}$  pin can be controlled by a host processor. As in the case of the VBAT pin (see above), the  $\overline{CE}$  pin should be connected to the host GPIO pin through as large a resistor as possible. The limitation on the resistor value is that the minimum V<sub>OH</sub> of the host GPIO pin less the drop across the resistor should be greater than V<sub>IH</sub> of the bq2431×  $\overline{CE}$  pin. The drop across the resistor is given by R<sub>CE</sub> × I<sub>IH</sub>.

The FAULT pin is an open-drain output that goes low during OV, OC, battery-OV, and OT events. If the application does not require monitoring of the FAULT pin, it can be left unconnected. But if the FAULT pin has to be monitored, it should be pulled high externally through  $R_{PU}$ , and connected to the host through  $R_{FAULT}$ .  $R_{FAULT}$  prevents damage to the host controller if the bq2431x fails (see above). The resistors should be of high value, in practice values between  $22k\Omega$  and  $100k\Omega$  should be sufficient.

### **Selection of Input and Output Bypass Capacitors**

The input capacitor  $C_{IN}$  in Figure 23 is for decoupling, and serves an important purpose. Whenever there is a step change downwards in the system load current, the inductance of the input cable causes the input voltage to spike up.  $C_{IN}$  prevents the input voltage from overshooting to dangerous levels. It is strongly recommended that a ceramic capacitor of at least 1µF be used at the input of the device. It should be located in close proximity to the IN pin.

 $C_{OUT}$  in Figure 23 is also important: If a very fast (< 1µs rise time) overvoltage transient occurs at the input, the current that charges  $C_{OUT}$  causes the device's current-limiting loop to kick in, reducing the gate-drive to FET Q1. This results in improved performance for input overvoltage protection.  $C_{OUT}$  should also be a ceramic capacitor of at least 1µF, located close to the OUT pin.  $C_{OUT}$  also serves as the input decoupling capacitor for the charging circuit downstream of the protection IC.

### **Powering Accessories**

In some applications, the equipment that the protection IC resides in may be required to provide power to an accessory (e.g. a cellphone may power a headset or an external memory card) through the same connector pins that are used by the adapter for charging. Figure 25 and Figure 26 illustrate typical charging and accessory-powering scenarios:



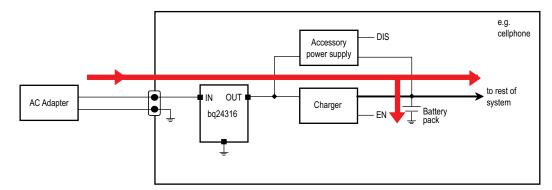


Figure 25. Charging - The Red Arrows Show the Direction of Current Flow

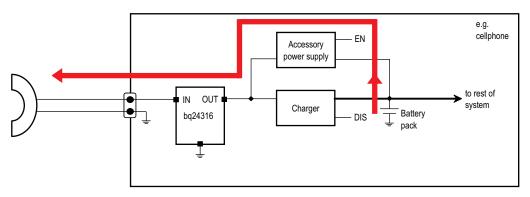
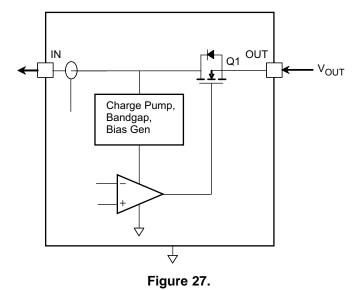


Figure 26. Powering an Accessory - The Red Arrows Show the Direction of Current Flow

In the second case, when power is being delivered to an accessory, the bq24314/bq24316 device is required to support current flow from the OUT pin to the IN pin.

If  $V_{OUT} > V_{UVLO} + 0.7V$ , FET Q1 is turned on, and the reverse current does not flow through the diode but through Q1. Q1 will then remain ON as long as  $V_{OUT} > V_{UVLO} - V_{HYS-UVLO} + R_{DS}ON*I_{ACCESSORY}$ . Within this voltage range, the reverse current capability is the same as the forward capability, 1.5A. It should be noted that there is no overcurrent protection in this direction.





### **PCB Layout Guidelines:**

- This device is a protection device, and is meant to protect down-stream circuitry from hazardous voltages. Potentially, high voltages may be applied to this IC. It has to be ensured that the edge-to-edge clearances of PCB traces satisfy the design rules for high voltages.
- The device uses SON packages with a PowerPAD<sup>™</sup>. For good thermal performance, the PowerPAD should be thermally coupled with the PCB ground plane. In most applications, this will require a copper pad directly under the IC. This copper pad should be connected to the ground plane with an array of thermal vias.
- C<sub>IN</sub> and C<sub>OUT</sub> should be located close to the IC. Other components like R<sub>ILIM</sub> and R<sub>BAT</sub> should also be located close to the IC.



# **Revision History**

CI	hanges from Revision B (September 2007) to Revision C Pa	age
•	Changed bq24314DSJ marking from preview to CBX	. 2
•	Changed bq24316DSJ marking from preview to BZC	. 2



11-Apr-2013

# **PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
BQ24314DSGR	ACTIVE	WSON	DSG	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBV	Samples
BQ24314DSGRG4	ACTIVE	WSON	DSG	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBV	Samples
BQ24314DSGT	ACTIVE	WSON	DSG	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBV	Samples
BQ24314DSGTG4	ACTIVE	WSON	DSG	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBV	Samples
BQ24314DSJR	ACTIVE	VSON	DSJ	12	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBX	Samples
BQ24314DSJRG4	ACTIVE	VSON	DSJ	12	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBX	Samples
BQ24314DSJT	ACTIVE	VSON	DSJ	12	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBX	Samples
BQ24314DSJTG4	ACTIVE	VSON	DSJ	12	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBX	Samples
BQ24316DSGR	ACTIVE	WSON	DSG	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBW	Samples
BQ24316DSGRG4	ACTIVE	WSON	DSG	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBW	Samples
BQ24316DSGT	ACTIVE	WSON	DSG	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBW	Samples
BQ24316DSGTG4	ACTIVE	WSON	DSG	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	CBW	Samples
BQ24316DSJR	ACTIVE	VSON	DSJ	12	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	BZC	Samples
BQ24316DSJRG4	ACTIVE	VSON	DSJ	12	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	BZC	Samples
BQ24316DSJT	ACTIVE	VSON	DSJ	12	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	BZC	Samples
BQ24316DSJTG4	ACTIVE	VSON	DSJ	12	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	BZC	Samples

<sup>(1)</sup> The marketing status values are defined as follows: **ACTIVE:** Product device recommended for new designs.





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LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect. NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design. PREVIEW: Device has been announced but is not in production. Samples may or may not be available. OBSOLETE: TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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# PACKAGE MATERIALS INFORMATION

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# TAPE AND REEL INFORMATION





# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
BQ24314DSGR	WSON	DSG	8	3000	179.0	8.4	2.2	2.2	1.2	4.0	8.0	Q2
BQ24314DSGT	WSON	DSG	8	250	179.0	8.4	2.2	2.2	1.2	4.0	8.0	Q2
BQ24314DSJR	VSON	DSJ	12	3000	330.0	12.4	3.3	4.3	1.1	8.0	12.0	Q1
BQ24314DSJT	VSON	DSJ	12	250	180.0	12.4	3.3	4.3	1.1	8.0	12.0	Q1
BQ24316DSGR	WSON	DSG	8	3000	179.0	8.4	2.2	2.2	1.2	4.0	8.0	Q2
BQ24316DSGT	WSON	DSG	8	250	179.0	8.4	2.2	2.2	1.2	4.0	8.0	Q2
BQ24316DSJR	VSON	DSJ	12	3000	330.0	12.4	3.3	4.3	1.1	8.0	12.0	Q1
BQ24316DSJT	VSON	DSJ	12	250	180.0	12.4	3.3	4.3	1.1	8.0	12.0	Q1

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# PACKAGE MATERIALS INFORMATION

26-Jan-2013



*All dimensions are nominal							
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
BQ24314DSGR	WSON	DSG	8	3000	195.0	200.0	45.0
BQ24314DSGT	WSON	DSG	8	250	195.0	200.0	45.0
BQ24314DSJR	VSON	DSJ	12	3000	367.0	367.0	35.0
BQ24314DSJT	VSON	DSJ	12	250	210.0	185.0	35.0
BQ24316DSGR	WSON	DSG	8	3000	195.0	200.0	45.0
BQ24316DSGT	WSON	DSG	8	250	195.0	200.0	45.0
BQ24316DSJR	VSON	DSJ	12	3000	367.0	367.0	35.0
BQ24316DSJT	VSON	DSJ	12	250	210.0	185.0	35.0

# **MECHANICAL DATA**



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-Leads (QFN) package configuration.

The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.

E. Falls within JEDEC MO-229.



# DSG (S-PWSON-N8)

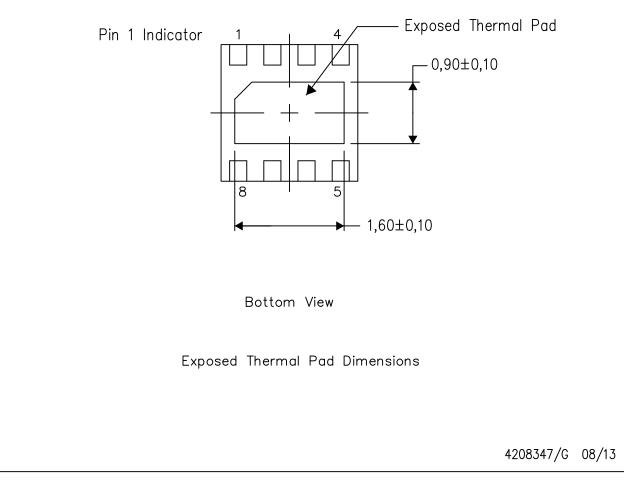
# PLASTIC SMALL OUTLINE NO-LEAD

#### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

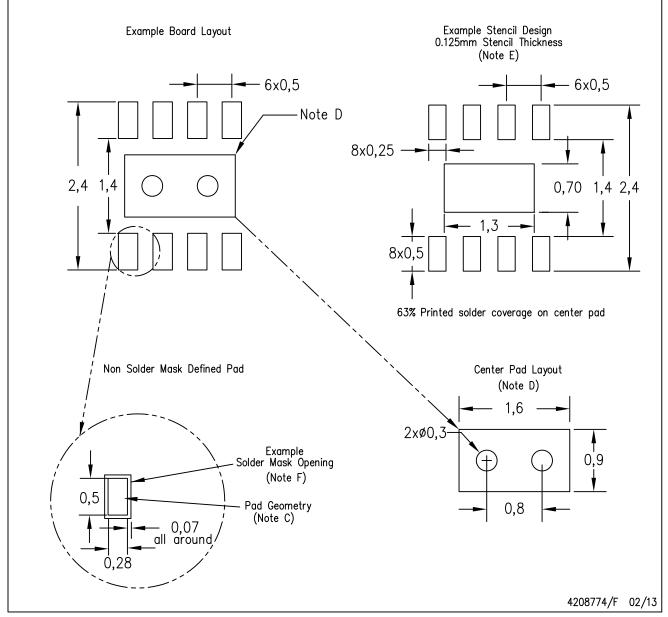


#### NOTE: All linear dimensions are in millimeters



DSG (S-PWSON-N8)

PLASTIC SMALL OUTLINE NO-LEAD

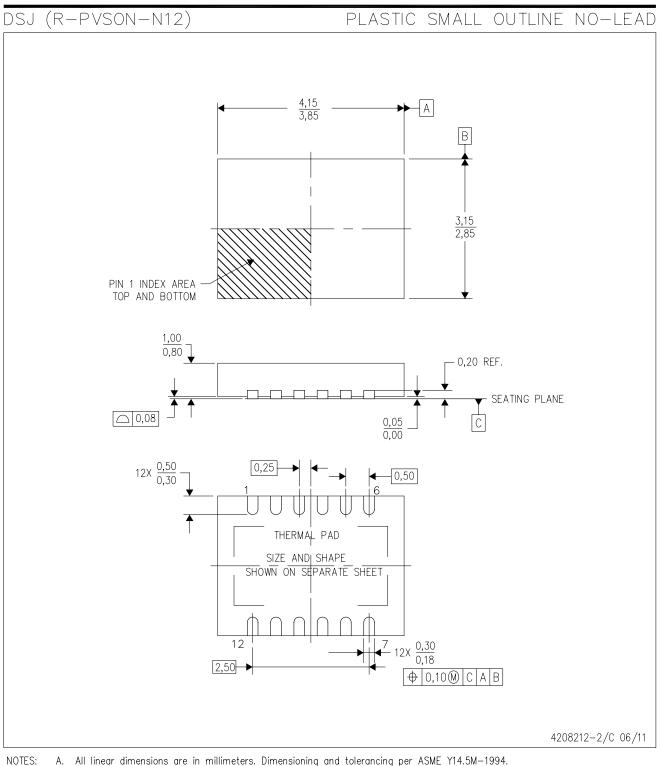


NOTES: A. All I

- A. All linear dimensions are in millimeters.B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN/SON PCB Attachment, Texas Instruments Literature No. <u>SLUA271</u>, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">http://www.ti.com</a>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for solder mask tolerances.

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# **MECHANICAL DATA**



A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- Β. This drawing is subject to change without notice.
- C. Quad Flatpack, No-Leads (QFN) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- F. Falls within JEDEC MO-229.



# THERMAL PAD MECHANICAL DATA



## THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. <u>SLUA271</u>. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

