INTEGRATED CIRCUITS



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HILIP

TDA3603

FEATURES

General

- One V_P state controlled regulator (regulator 2)
- Regulator 2, reset and ignition buffer operate during load dump and thermal shutdown
- One control pin for switching regulator 1 and the power switch
- Supply voltage range of -18 to +50 V (operating from 9.75 V)
- Low reverse current of regulator 2
- Low quiescent current (when regulator 1, power switch and ignition input are switched off, standby)
- Ignition input/output
- Reset output
- High ripple rejection
- Power switch.

Protections

- Reverse polarity safe (down to -18 V without high reverse current)
- Able to withstand voltages up to 18 V at the outputs (supply line may be shortened)
- ESD protected on all pins
- Thermal protection
- · Load dump protection
- Foldback current limit protection for regulators 1 and 2
- Delayed second current limit protection for the power switch
- The regulator outputs and the power switch are DC short-circuited safe to ground and V_P.

ORDERING INFORMATION

GENERAL DESCRIPTION

The TDA3603 is a multiple output voltage regulator with a power switch, intended for use in car radios with or without a microcontroller.

It contains one fixed voltage regulator with a foldback current protection (regulator 1) and one fixed voltage regulator (regulator 2), intended to supply a microcontroller, that also operates during load dump and thermal shutdown.

There is a power switch with protections, operated by the enable input.

The reset and ignition outputs can be used to interface by the microcontroller. The reset signal can be used to call up the microcontroller and the ignition output indicates ignition voltage available.

The supply pin can withstand load dump pulses and negative supply voltages.

Regulator 2 will be switched on at a supply voltage >6.5 V and off at a voltage of regulator 2 <1.9 V.

		PACKAGE				
	NAME	DESCRIPTION	VERSION			
TDA3603	SIL9MPF	SIL9MPF plastic single in-line medium power package with fin; 9 leads				
TDA3603P	HDIP18 plastic heat-dissipating dual in-line package; 18 leads		SOT398-1			

TDA3603

QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply						
V _P	supply voltage					
	operating		9.75	14.4	25	V
	regulator 2 on	note 1	2.4	14.4	25	V
	jump start	t ≤ 10 minutes	-	-	30	V
	load dump protection	during 50 ms; $t_r \ge 2.5$ ms	_	-	50	V
lq	total quiescent current	standby mode	-	400	500	μΑ
T _{vj}	operating virtual junction temperature		-	_	150	°C
Voltage regul	lators					
V _{REG1}	output voltage regulator 1	$0.5 \text{ mA} \le I_{REG1} \le 300 \text{ mA}$	8.65	9.0	9.35	V
V _{REG2}	output voltage regulator 2	0.5 mA \leq I _{REG2} \leq 50 mA; V _P = 14.4 V	4.8	5.0	5.2	V
V _{REGd1}	drop-out voltage regulator 1	I _{REG1} = 0.3 A; note 2	-	_	0.5	V
Power switch	1					
V _{swd}	drop-out voltage	I _{sw} = 0.3 A; note 3	_	_	0.9	V
I _M	peak current	$t \le 10 \text{ ms}$	1.4	_	-	A

Notes

1. Minimum operating voltage, only if V_P has exceeded 6.5 V.

2. The drop-out voltage of regulator 1 is measured between V_P and REG1.

3. The drop-out voltage of the power switch is measured between V_{P} and $V_{\text{sw}}.$

BLOCK DIAGRAM



PINNING

SYMDOL	PIN		DECODIDITION	
STMBOL	SOT110-1	SOT398-1	DESCRIPTION	
V _P	1	1	supply voltage	
REG1	2	2	regulator 1 output	
RES	3	3	reset output voltage (+5 V)	
V _{en}	4	4	enable input voltage	
V _{O(ig)}	5	5	ignition output voltage	
GND	6	6	ground (0 V)	
REG2	7	7	regulator 2 output	
V _{I(ig)}	8	8	ignition input voltage	
V _{sw}	9	9	power switch output voltage	
i.c.	_	10 to 18	can be connected to a heat spreader	



FUNCTIONAL DESCRIPTION

The TDA3603 is a multiple output voltage regulator with a power switch, intended for use in car radios with or without a microcontroller. Because of low-voltage operation of the car radio, low-voltage drop regulators are used.

Regulator 2 will switch on when the supply voltage exceeds 6.5 V for the first time and will switch off again when the output voltage of regulator 2 is below 1.9 V (this is below an engine start). When regulator 2 is switched on and the output voltage of this regulator is within its voltage range, the reset output will be enabled (reset will go HIGH via a pull-up resistor) to generate a reset to the microcontroller. The reset cycles can be extended by an external capacitor at the reset output (pin 3). The start-up feature is built-in to ensure a smooth start-up of the microcontroller at first connection, without uncontrolled switching of regulator 2 during the start-up sequence.

When both regulator 2 and the supply voltage ($V_P > 4.5 V$) are available, regulator 1 and the switch can be operated by an enable input (pin 4).

All output pins are fully protected. The regulators are protected against load dump (regulator 1 will switch off at supply voltages higher than 25 V) and short-circuit (foldback current protection).

The switch contains a current protection which is delayed for \geq 10 ms (in short-circuit condition). During this time the current is limited to 1.4 A (V_P \leq 18 V).

At supply voltages over 16.9 V the switch is clamped at 15.0 V (to avoid externally connected circuitry being damaged by an overvoltage) and the switch will switch off at load dump.

Interfacing with the microcontroller can be accomplished by an ignition Schmitt trigger and ignition output buffer, (simple full/semi on/off logic applications).

The total timing of a semi on/off logic set is shown Fig.4.



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LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V _P	supply voltage				
	operating		-	25	V
	jump start	t ≤ 10 minutes	-	30	V
	load dump protection	during 50 ms; $t_r \ge 2.5$ ms	-	50	V
VP	reverse battery voltage	non-operating	-	–18	V
V _{ppi}	positive pulse voltage at ignition buffer	$V_{P} = 14.4 \text{ V}; \text{ R}_{I} = 1 \text{ k}\Omega$	-	50	V
V _{npi}	negative pulse voltage at ignition buffer	$V_{P} = 14.4 \text{ V}; \text{ R}_{I} = 1 \text{ k}\Omega$	_	-100	V
T _{stg}	storage temperature	non-operating	-55	+150	°C
T _{vj}	operating virtual junction temperature		-40	+150	°C
P _{tot}	total power dissipation				
	SOT110-1		-	10.4	W
	SOT398-1		-	8.3	W

THERMAL CHARACTERISTICS

SYMBOL	TYPE NUMBER	PARAMETER	VALUE	UNIT
R _{th j-c}	TDA3603	thermal resistance from junction to case	12	K/W
R _{th j-p}	TDA3603P	thermal resistance from junction to pins	15	K/W

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CHARACTERISTICS

 V_P = 14.4 V; T_{amb} = 25 °C; see Fig.7; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply		•		•		
V _P	supply voltage					
	operating		9.75	14.4	25	V
	regulator 2 on	note 1	2.4	14.4	25	V
	jump start	t ≤ 10 minutes	_	-	30	V
	load dump protection	during 50 ms; $t_r \ge 2.5$ ms	-	-	50	V
lq	quiescent current	V _P = 12.4 V; note 2	-	400	500	μA
		V _P = 14.4 V; note 2	_	420	-	μA
Schmitt trig	gger power supply for the po	wer switch				
V _{thr}	rising voltage threshold		4.0	4.5	5.0	V
V _{thf}	falling voltage threshold		3.5	4.0	4.5	V
V _{hys}	hysteresis		_	0.5	_	V
Schmitt trig	gger for regulator 1					
V _{thr}	rising voltage threshold		4.0	4.5	5.0	V
V _{thf}	falling voltage threshold		3.5	4.0	4.5	V
V _{hys}	hysteresis		_	0.5	-	V
Schmitt trig	gger power supply for regulat	tor 2				
V _{thr}	rising voltage threshold		6.0	6.5	7.1	V
V _{thf}	falling voltage threshold		1.7	1.9	2.2	V
V _{hys}	hysteresis		-	4.7	-	V
Schmitt trig	gger for enable input					
V _{thr}	rising voltage threshold		1.7	2.2	2.7	V
V _{thf}	falling voltage threshold		1.5	2.0	2.5	V
V _{hys}	hysteresis		_	0.2	-	V
Schmitt trig	gger for reset buffer		·			
V _{r(REG2)}	rising voltage of regulator 2	note 3	-	V _{REG2} - 0.15	_	V
V _{f(REG2)}	falling voltage of regulator 2	note 3	_	$V_{REG2} - 0.25$	-	V
V _{spread}	voltage spread on tracking	note 4	_	10	-	mV
Schmitt trigger for ignition buffer						
V _{thr}	rising voltage threshold		1.7	2.2	2.7	V
V _{thf}	falling voltage threshold		1.5	2.0	2.5	V
V _{hys}	hysteresis		-	0.2	_	V
Reset buffe	er					
l _{sink}	LOW-level sink current	$V_{RES} \le 0.8 V$	15	20	_	mA
I _{leak}	leakage current	V _P = 14.4 V; V _{RES} = 5 V	25	50	100	μA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Ignition bu	Ignition buffer					1
V _{OL}	LOW-level output voltage	$I_{OL} = 0 \text{ mA}$	0	0.2	0.8	V
V _{OH}	HIGH-level output voltage	note 5	_	5.0	5.2	V
I _{OL}	LOW-level output current	$V_{OL} \le 0.8 \text{ V}$	0.3	0.8	_	mA
I _{OH}	HIGH-level output current	V _{OH} ≥ 3 V	0.3	2.0	-	mA
Regulator '	I; note 6					!
V _{REG1}	output voltage off		_	1	400	mV
V _{REG1}	output voltage	$0.5 \text{ mA} \le I_{\text{REG1}} \le 300 \text{ mA}$	8.65	9.0	9.35	V
		$10 \text{ V} \le \text{V}_{\text{P}} \le 18 \text{ V}$	8.65	9.0	9.35	V
ΔV_{REG1}	line regulation	$10 \text{ V} \le \text{V}_{P} \le 18 \text{ V}$	-	-	50	mV
ΔV_{REGL1}	load regulation	$0.5 \text{ mA} \le I_{\text{REG1}} \le 300 \text{ mA}$	-	-	70	mV
SVRR1	supply voltage ripple rejection	f _i = 200 Hz; V _I = 2 V (p-p)	60	-	-	dB
V _{REGd1}	drop-out voltage	I _{REG1} = 300 mA; note 7	-	0.4	0.5	V
I _{REGm1}	current limit	V _{REG1} > 7 V; note 8	0.45	-	1.2	А
I _{REGsc1}	short-circuit current	$R_L \le 0.5 \Omega$; note 9	50	300	-	mA
α _{ct}	cross talk	note 10	-	50	-	dB
Regulator 2	2; note 11					
V _{REG2}	output voltage	$0.5 \text{ mA} \le I_{\text{REG2}} \le 50 \text{ mA}$	4.8	5.0	5.2	V
		$7 \text{ V} \le \text{V}_{P} \le 18 \text{ V}$	4.8	5.0	5.2	V
		$18 \text{ V} \le \text{V}_{\text{P}} \le 50 \text{ V}$	4.75	5.0	5.25	V
ΔV_{REG2}	line regulation	$7 \text{ V} \le \text{V}_{P} \le 18 \text{ V}$	-	-	50	mV
ΔV_{REGL2}	load regulation	$0.5 \text{ mA} \le I_{\text{REG1}} \le 30 \text{ mA}$	-	-	50	mV
SVRR2	supply voltage ripple rejection	f _i = 200 Hz; V _I = 2 V (p-p)	60	-	-	dB
V _{REGd2}	drop-out voltage	I _{REG2} = 30 mA; note 12	-	0.3	0.4	V
I _{REGm2}	current limit	V _{REG2} > 4.5 V; note 8	0.1	-	0.5	A
I _{REGsc2}	short-circuit current	$R_L \le 0.5 \Omega$; note 9	-	50	-	mA
α_{ct}	cross talk	note 13	-	50	_	dB
Power swit	ch					
V _{swd}	drop-out voltage	I _{sw} = 0.3 A; note 14	-	0.4	0.9	V
I _{swcc}	continuous current		0.5	_	-	A
V _{swcl}	clamping voltage	V _P ≥ 16.9 V	-	15.0	16.2	V
I _M	peak current	t ≤ 10 ms	1.4	-	-	A
V _{swfb}	fly back voltage behaviour	$I_{sw} = -200 \text{ mA}, V_{P} = 9 \text{ V}$	_	_	20	V
I _{lim(sw)}	current limit	V _P = 14.4 V; V _{sw} < 1.5 V; note 8	0.6	-	1.0	A

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Notes to the characteristics

- 1. Minimum operating voltage, only if V_P has exceeded 6.5 V.
- 2. Enable and ignition inputs are low and regulator 2 is unloaded.
- 3. Voltage drop due to load condition.
- 4. The spread on tracking is one sigma value.
- 5. Ignition output voltage will be less than or equal to the output voltage of regulator 2.
- 6. I_{REG1} = 5 mA.
- 7. The drop-out voltage of regulator 1 is measured between V_P and REG1.
- 8. At current limit, I_{REGm} is held constant (see Fig.5).
- 9. The foldback current protection limits the dissipated power at short-circuit (see Figs 5 and 6).
- 10. The cross talk of regulator 1 is measured with an I_{REG2} = 0.5 mA up to 30 mA with an input frequency of f_i = 100 kHz.
- 11. I_{REG2} = 5 mA.
- 12. The drop-out voltage of regulator 2 is measured between V_{P} and REG2.
- 13. The cross talk of regulator 2 is measured with an I_{REG1} = 0.5 mA up to 100 mA with an input frequency of f_i = 100 kHz.
- 14. The drop-out voltage of the power switch is measured between V_{P} and $V_{\mathsf{sw}}.$

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TEST AND APPLICATION INFORMATION



Noise information

The noise at the output of the regulators depends on the bandwidth of the regulators, which can be adjusted by the output capacitors. Table 1 shows the noise figures.

Although stability is guaranteed when C_L is higher than 10 μ F (over temperature range) with tan (ϕ) = 1 in the frequency range 1 to 10 kHz, however, for low noise, a 47 μ F load capacitor is required. When electrolytic capacitors are used, the capacitor value will decrease and the ESR will increase much at low temperatures. To avoid oscillation a normal capacitor of 220 nF can be placed in parallel with this electrolytic capacitor.

The noise on the supply line depends on the value of the supply capacitor and is caused by a current noise (output noise of the regulators is translated into a current noise by the output capacitors). When a high frequency capacitor of 220 nF with an electrolytic capacitor of 100 μ F in parallel is placed directly over pins 1 and 6 (supply and ground) the noise is minimized.

Table 1Noise figures

REGULATOR	NOISE (μV) ⁽¹⁾	OUTPUT CAPACITOR (μF)
	180	10
1	100	47
	80	100
	120	10
2	70	47
	70	100

Note

1. Bandwidth of 100 kHz.

SHORT CIRCUIT BEHAVIOUR OF POWER SWITCH

The short circuit behaviour of the switch with large inductive loads (switch output goes out of the radio) can be improved by replacing C2 (see Fig.7) by a larger electrolytic capacitor of 10 μ F/16 V. When the temperature protection of the switch becomes active, due to a short circuit of the switch, the behaviour will be improved.

When the switch is clamped an minimum output capacitor of 10 μF is needed.

The power switch is not protected against 'loss of ground' condition (= short of the switch to ground with floating ground pin of the TDA3603 itself). A 'loss of ground' situation can in practice only occur when the switch output goes outside the car-radio box. There is an application solution to protect against 'loss of ground' (see Fig.8).

It is advisable to limit the dissipation at short circuit condition by monitoring the output of the power switch. The microprocessor can switch of the power switch when the switch was enabled and the switch output remains low due to a short circuit condition.



PACKAGE OUTLINES



TDA3603



HDIP18: plastic heat-dissipating dual in-line package; 18 leads

TDA3603

SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

Soldering by dipping or by wave

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact

time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{stg\,max}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

Repairing soldered joints

Apply a low voltage soldering iron (less than 24 V) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than $300 \,^{\circ}$ C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 °C, contact may be up to 5 seconds.

DEFINITIONS

Data sheet status			
Objective specification	bjective specification This data sheet contains target or goal specifications for product development.		
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.		
Product specification	This data sheet contains final product specifications.		
Limiting values			
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.			
Application information			
Where application information is given, it is advisory and does not form part of the specification.			

LIFE SUPPORT APPLICATIONS

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